

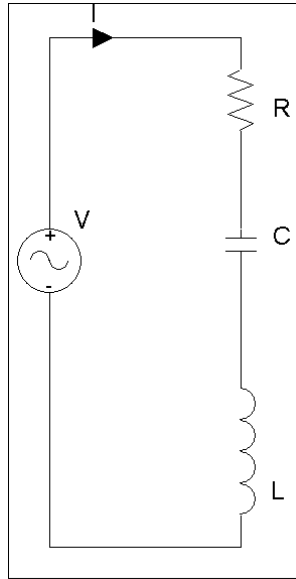


ELECTRICAL MACHINES-I

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Review of 1-Φ AC Circuit Fundamentals



Series RLC circuit

$$I \angle -\theta = \frac{V \angle 0}{R + j\omega L - \frac{j}{\omega C}}$$

$$\theta = \tan^{-1} \left[\frac{\left(\omega L - \frac{1}{\omega C} \right)}{R} \right]$$

Review of 1-Φ AC Circuit Fundamentals(1)

$$I \angle -\theta = I \cos \theta - j I \sin \theta$$

$$\text{Power factor} = \cos \theta = \frac{\text{Real Power}}{\text{Apparent Power}}$$

Apparent Power = VI (multiply the rms value of input voltage and current (ignore phase angle))

Real Power = $I^2 R$ (square of the rms current flowing through the resistor times the resistor (ignore phase angle))

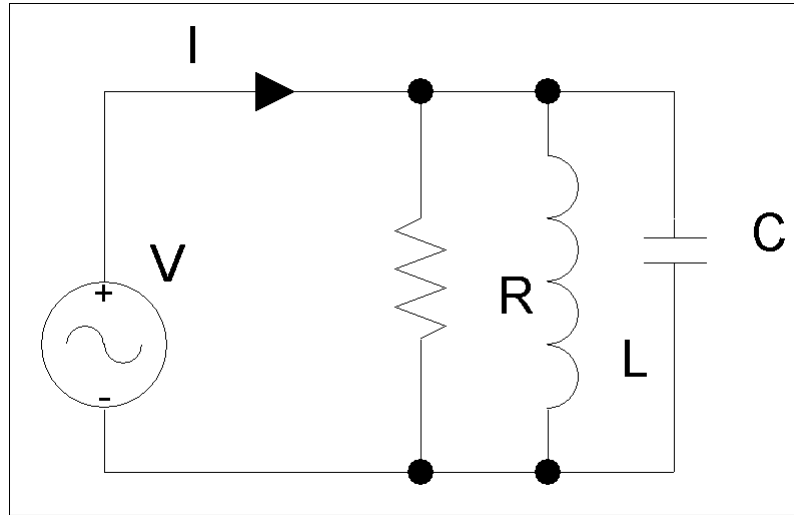
Series Resonance occurs when $\omega L - \frac{1}{\omega C} = 0$

$$\omega = \frac{1}{\sqrt{LC}}$$

$$I = \frac{V \angle 0}{R} \quad \text{is maximum in this case}$$

Transformer

Review of 1-Φ AC Circuit Fundamentals(3)



Parallel RLC circuit

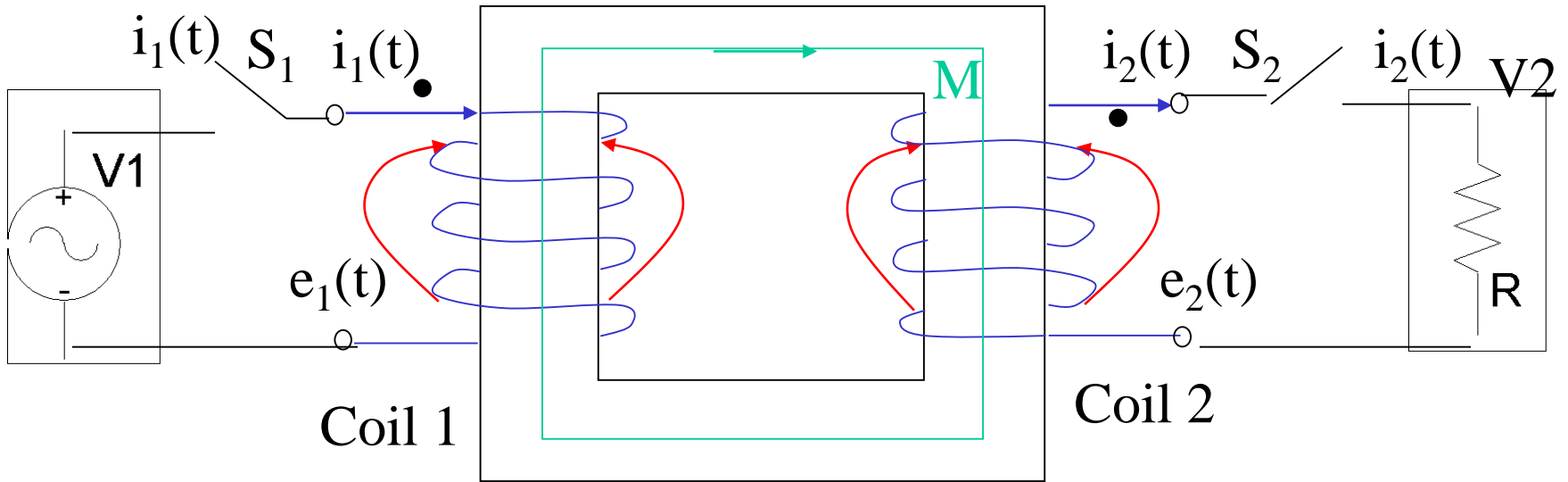
Parallel Resonance occurs when $\omega L - \frac{1}{\omega C} = 0$

$$\omega = \frac{1}{\sqrt{LC}}$$

$$I = \frac{V \angle 0}{R} \quad \text{is minimum in this case}$$

Transformer

The Transformer



(Primary has N_1 turns)

(Secondary has N_2 turns)

The Transformer(2)

- The source side is called Primary
- The load side is called Secondary
- Ideally
 1. The resistance of the coils are zero.
 2. The relative permeability of the core is infinite.
 3. Zero core or iron loss.
 4. Zero leakage flux

The Transformer(2)

i) Switch ' S_1 ' is closed and ' S_2 ' is open at $t=0$

The core does not have a flux at $t=0$

We will now prove the following on the greenboard:

The voltage induced across each coil is proportional to its number of turns.

The Transformer(3)

ii) Switch 'S₂' is now closed

A current now starts to flow in resistance R. This current is $i_2(t)$ (flows out of the dotted terminal).

$$i_2(t) = \frac{e_2(t)}{R} = \frac{V_2(t)}{R}$$

Thus a MMF $N_2 i_2(t)$ is applied to the magnetic circuit. This will immediately make a current $i_1(t)$ flow into the dot of the primary side, so that $N_1 i_1(t)$ opposes $N_2 i_2(t)$ and the original flux in the core remains constant. Otherwise, $N_2 i_2(t)$ would make the core flux change drastically and the balance between V_1 and $e_1(t)$ will be disturbed.

The Transformer(3)

We will now prove the following on the greenboard:

- 1) The current induced in each coil is inversely proportional to its number of turns.
- 2) Instantaneous input power to the transformer = Instantaneous output power from the transformer.

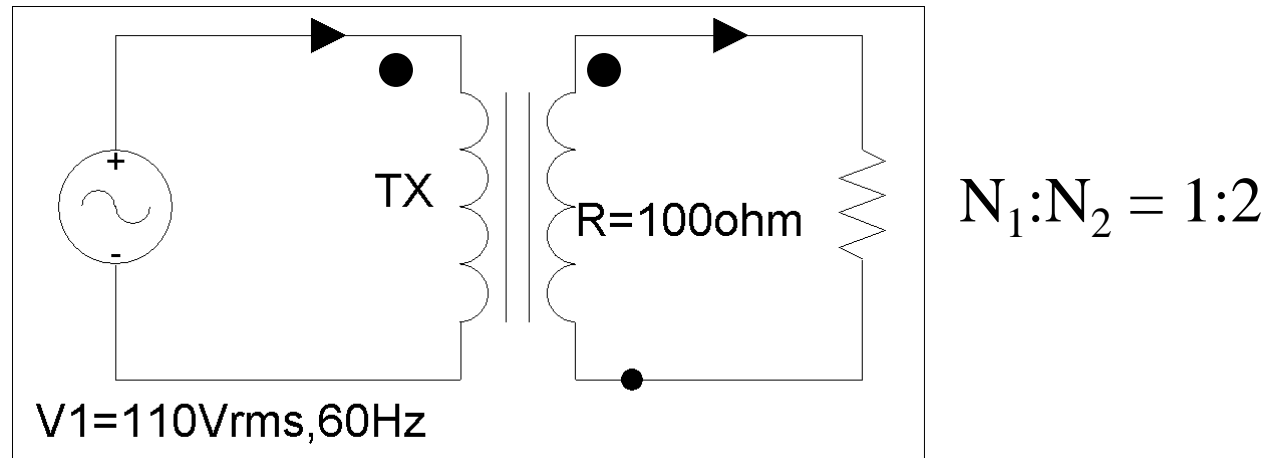
The Transformer(3)

Observation: It was shown that the flux in the core is $\Phi_m \sin(\omega t)$. Since the permeability of the core is infinite ideally zero current can produce this flux! In actuality, a current I_m , known as magnetizing current is required to setup the flux in the transformer. This current is within 5% of the full load current in a well designed transformer.

$$I_m = \frac{V_{1rms}}{\omega L_1}; L_1 = \frac{N_1^2}{\mathfrak{R}}$$

L_1 is the primary side self inductance.

Transformer Example(1)



- i) Find I_1, I_2 in the above transformer. Neglect magnetizing current.
- ii) What is the reflected (referred) load impedance on the primary side
- iii) If the resistance is replaced by a) 100 mH inductor b) $10\mu\text{F}$ capacitance; what will be the reflected load impedance on the primary side?

Transformer Example(1)

Solution on greenboard

Polarity (dot) convention

Terminals of different windings are of same **polarity** if currents entering (or leaving) them produce flux in the same direction in the core.

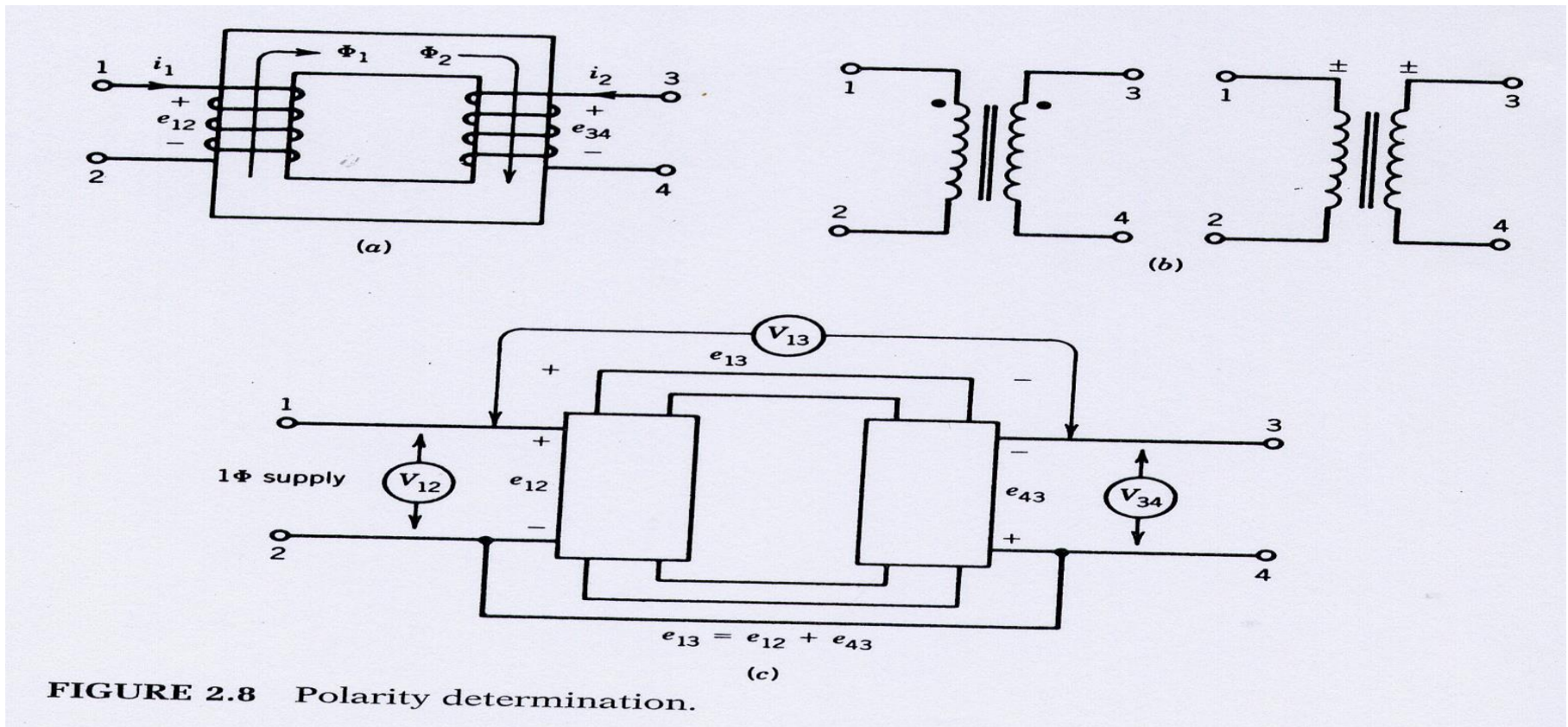


FIGURE 2.8 Polarity determination.

How to check polarity?

- 1) Measure e_{12} and e_{34}
- 2) Connect 2 and 4 and measure e_{13}
- 3) If $e_{13} = e_{12} + e_{34}$, 1 and 4 have same polarity
- 4) If $e_{13} = e_{12} - e_{34}$, 1 and 4 have different polarity

Parallel operation of transformers

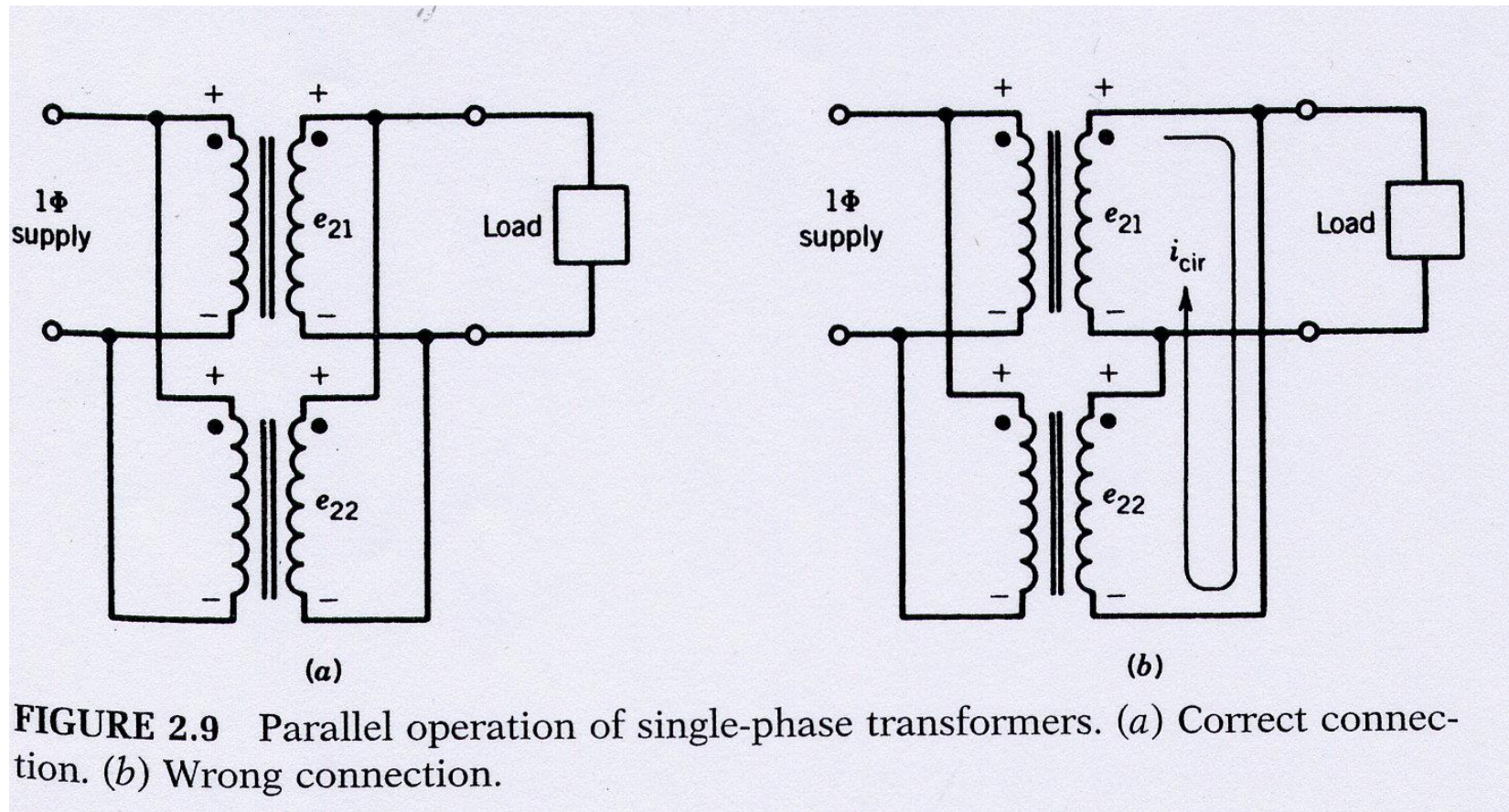
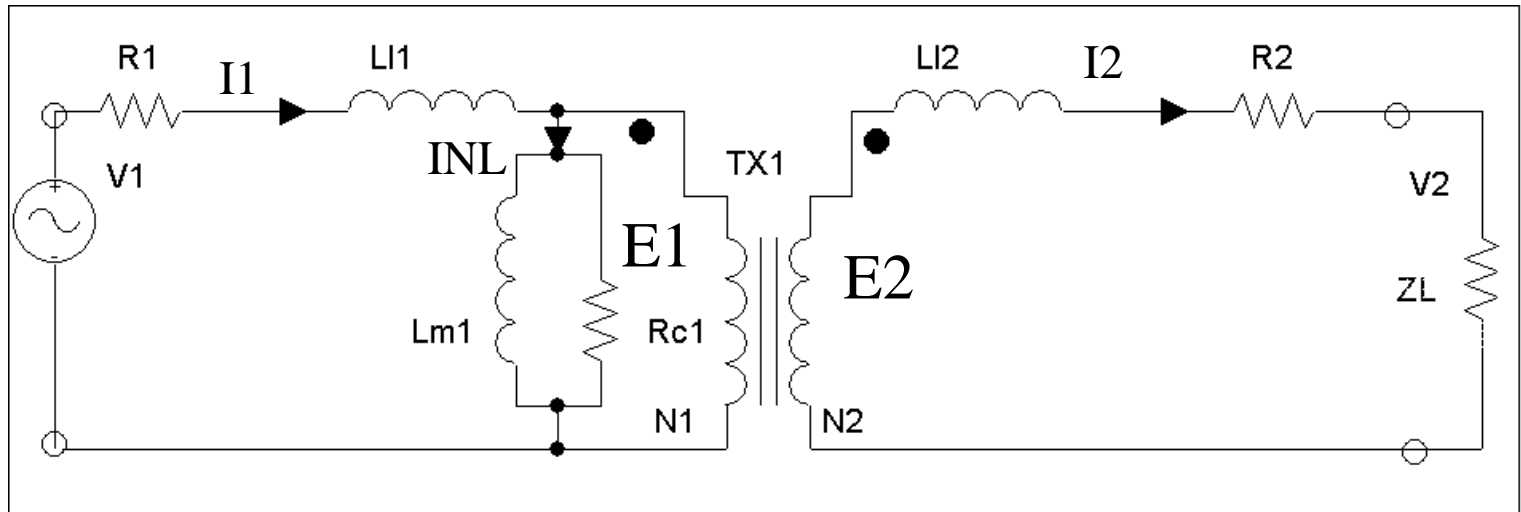


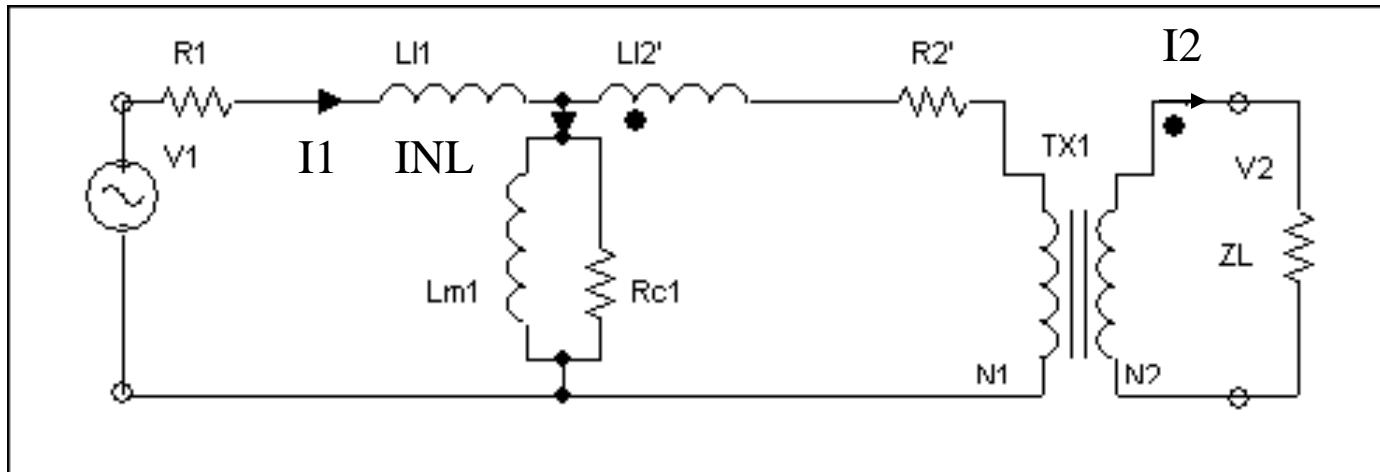
FIGURE 2.9 Parallel operation of single-phase transformers. (a) Correct connection. (b) Wrong connection.

Wrong connections give circulating between the windings that can destroy transformers.

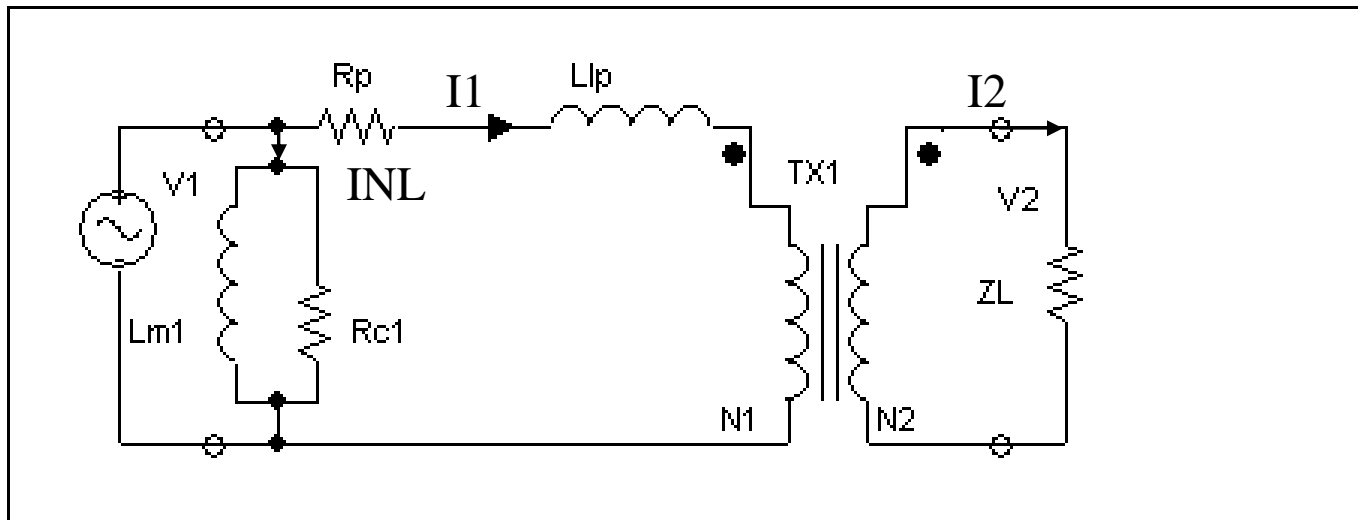
Transformer Equivalent circuit (1)



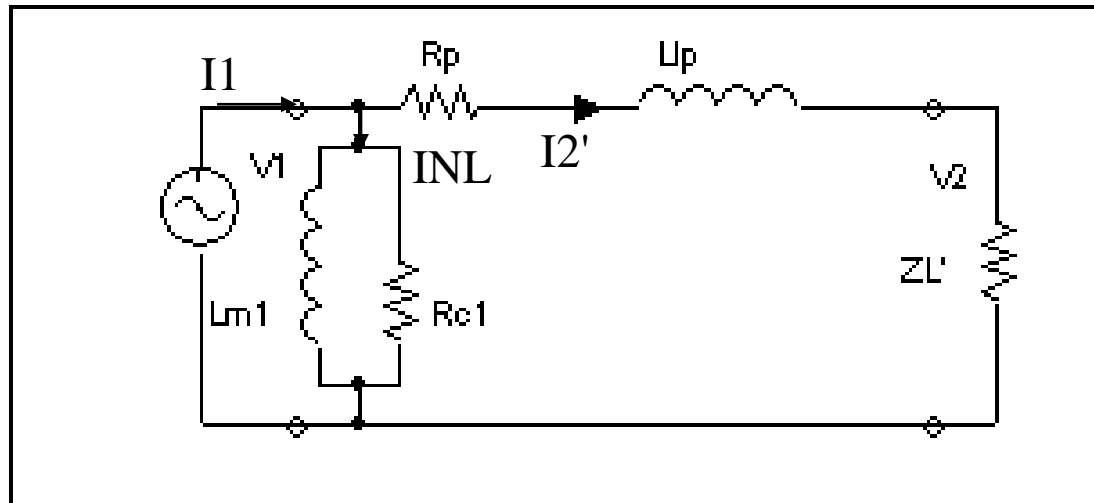
Transformer Equivalent circuit (2)



Transformer Equivalent circuit (3)



Transformer Equivalent circuit (4)



Open circuit Test

- It is used to determine L_{m1} (X_{m1}) and R_{c1}
- Usually performed on the low voltage side
- The test is performed at rated voltage and frequency under no load

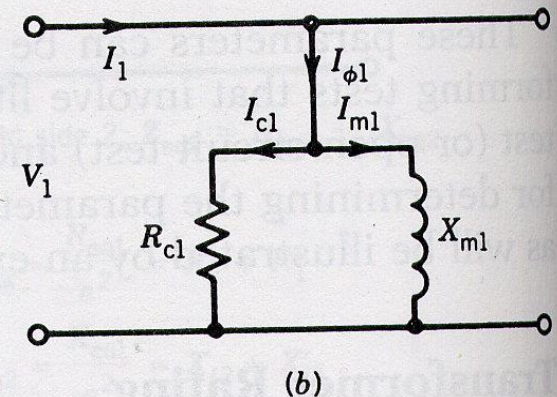
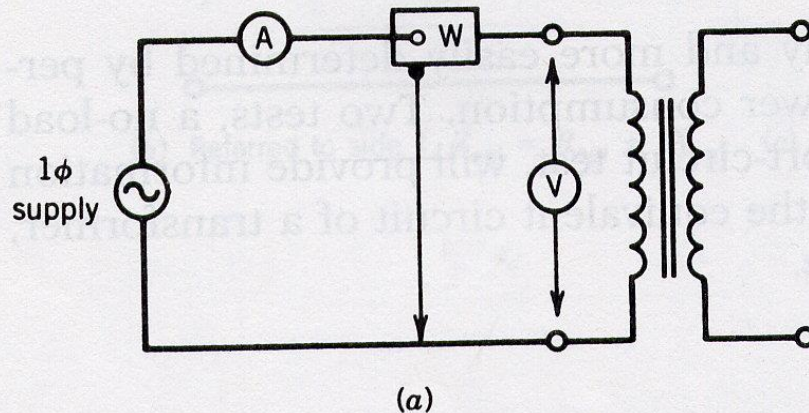
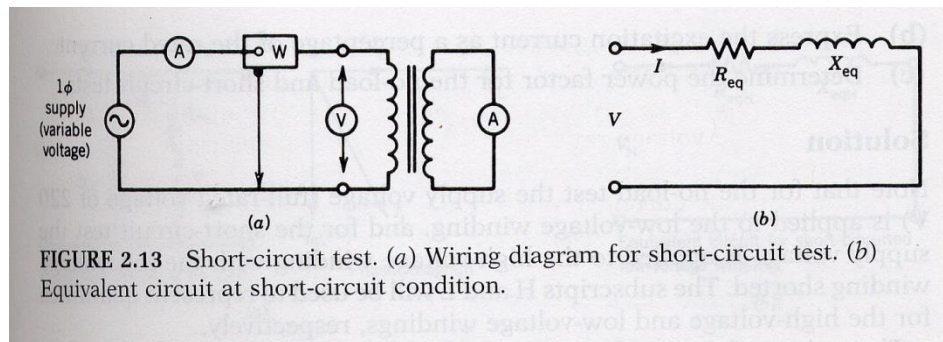


FIGURE 2.12 No-load (or open-circuit) test. (a) Wiring diagram for open-circuit test. (b) Equivalent circuit under open circuit.

Short circuit Test

- It is used to determine L_p (X_{eq}) and R_p (R_{eq})
- Usually performed on the high voltage side
- This test is performed at *reduced* voltage and rated frequency with the output of the low voltage winding short circuited such that rated current flows on the high voltage side.



Transformer Regulation

- Loading changes the output voltage of a transformer. Transformer regulation is the measure of such a deviation.

Definition of % Regulation

$$= \frac{|V_{\text{no-load}}| - |V_{\text{load}}|}{|V_{\text{load}}|} * 100$$

$V_{\text{no-load}}$ = RMS voltage across the load terminals without load

V_{load} = RMS voltage across the load terminals with a specified load

Maximum Transformer Regulation

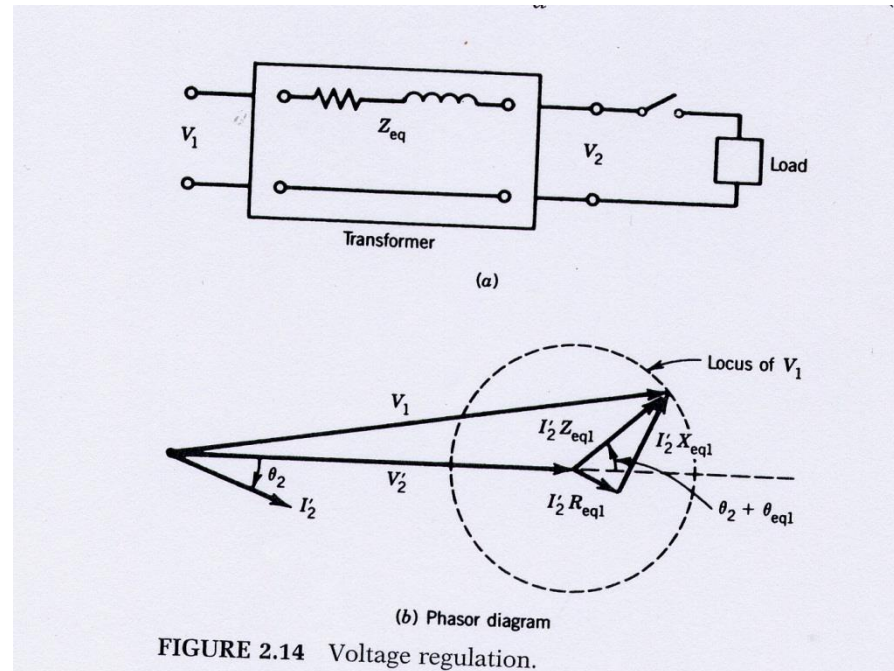


FIGURE 2.14 Voltage regulation.

$$V_1 = V_2' \angle 0^\circ + I_2' \angle \theta_2 \cdot Z_{eq1} \angle \theta_{eq1}$$

Clearly V_1 is maximum when

$$\theta_2 + \theta_{eq1} = 0; \text{ or } \theta_2 = -\theta_{eq1}$$

Transformer Losses and Efficiency

- Transformer Losses

- Core/Iron Loss = V_1^2 / R_{c1}

- Copper Loss = $I_1^2 R_1 + I_2^2 R_2$

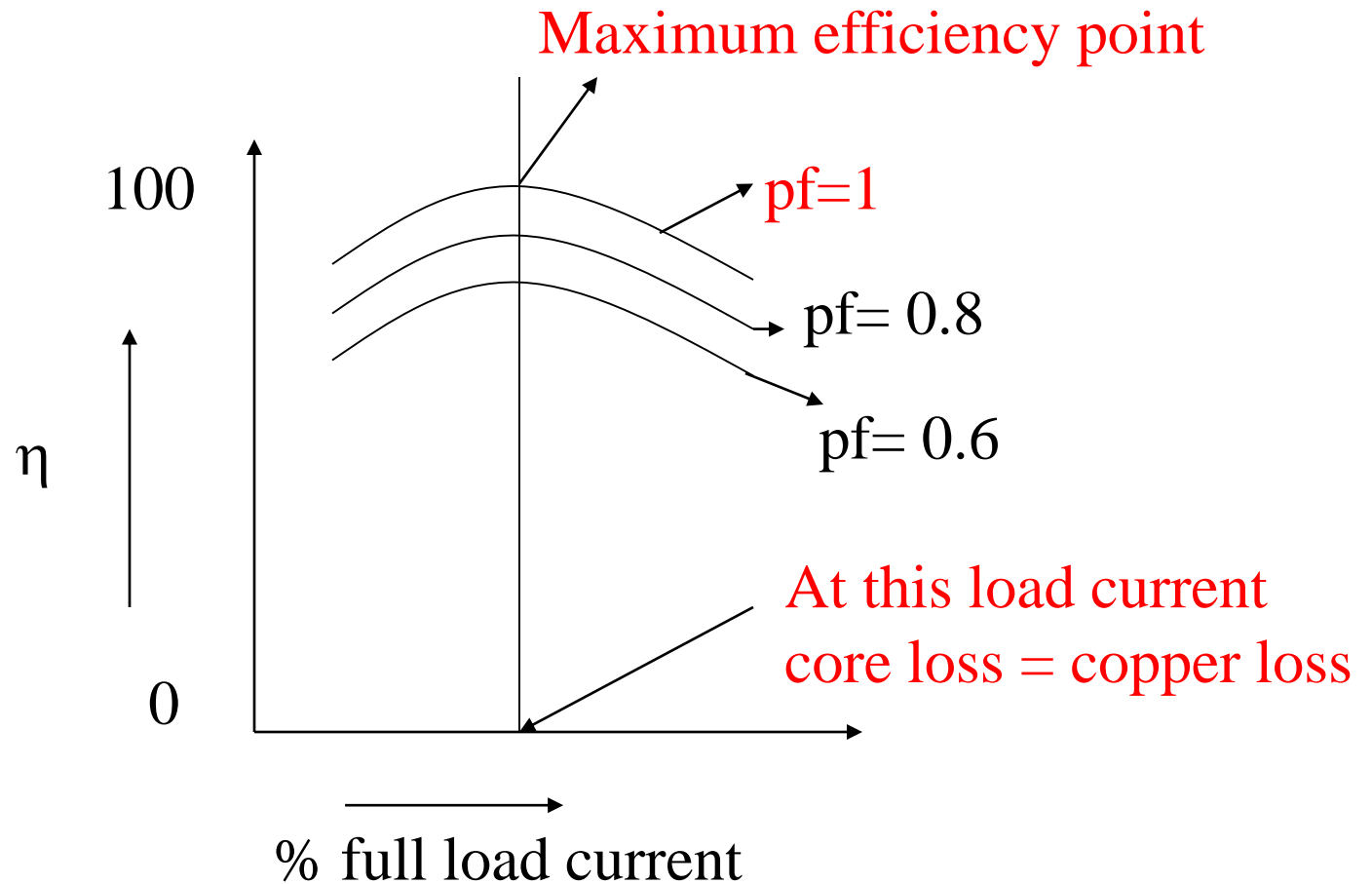
Definition of % efficiency

$$\begin{aligned} &= \frac{V_2 I_2 \cos \theta_2}{\text{Losses} + V_2 I_2 \cos \theta_2} * 100 \\ &= \frac{V_2 I_2 \cos \theta_2}{V_1^2 / R_{c1} + I_1^2 R_1 + I_2^2 R_2 + V_2 I_2 \cos \theta_2} * 100 \\ &= \frac{V_2 I_2 \cos \theta_2}{V_1^2 / R_{c1} + I_2^2 R_{eq2} + V_2 I_2 \cos \theta_2} * 100 \\ &\quad \cos \theta_2 = \text{load power factor} \end{aligned}$$

Maximum Transformer Efficiency

The efficiency varies as with respect to 2 independent quantities namely, current and power factor

- Thus at any particular power factor, the efficiency is maximum if **core loss = copper loss**. This can be obtained by differentiating the expression of efficiency with respect to I_2 assuming power factor, and all the voltages constant.
- At any particular I_2 maximum efficiency happens at **unity power factor**. This can be obtained by differentiating the expression of efficiency with respect to power factor, and assuming I_2 and all the voltages constant.
- Maximum efficiency happens when both these conditions are satisfied.



Another Transformer Example

The following are the open circuit and short circuit test data of a single phase, 10 kVA, 2200/220V, 60 Hz transformer

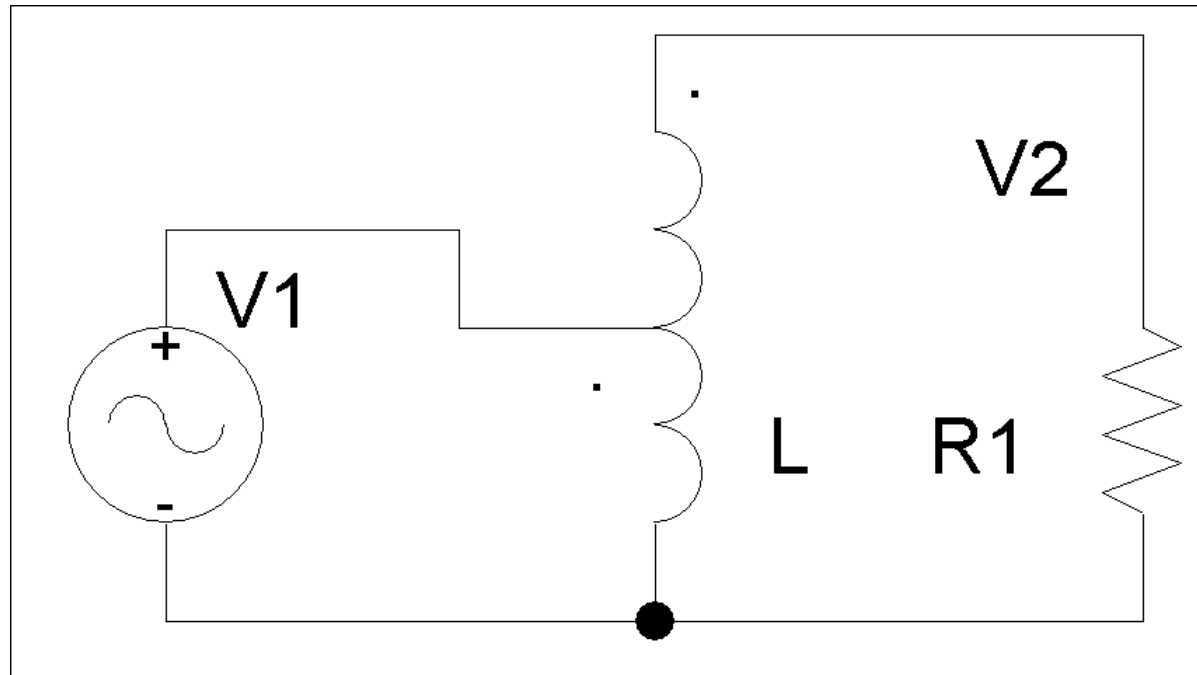
	O/C Test (HV side Open)	S/C Test (LV side Shorted)
Voltmeter	220V	150V
Ammeter	2.5A	4.55A
Wattmeter	100W	215W

- i) Find the equivalent circuit with respect to HV and LV side
- ii) Find the efficiency and regulation of the transformer when supplying rated load at 0.8 pf lag.
- iii) Maximum efficiency and regulation.

Transformer Example(2)

Solution on greenboard

Autotransformer



- Primary and secondary on the same winding. Therefore there is no galvanic isolation.

Features of Autotransformer

- ✓ Lower leakage
- ✓ Lower losses
- ✓ Lower magnetizing current
- ✓ Increase kVA rating
- × No galvanic Isolation

Autotransformer Theory and Example

Explained and worked out on Greenboard

Review of balanced three phase circuits

- Two possible configurations: Star (Y) and delta (Δ)
- Star has neutral, delta does not

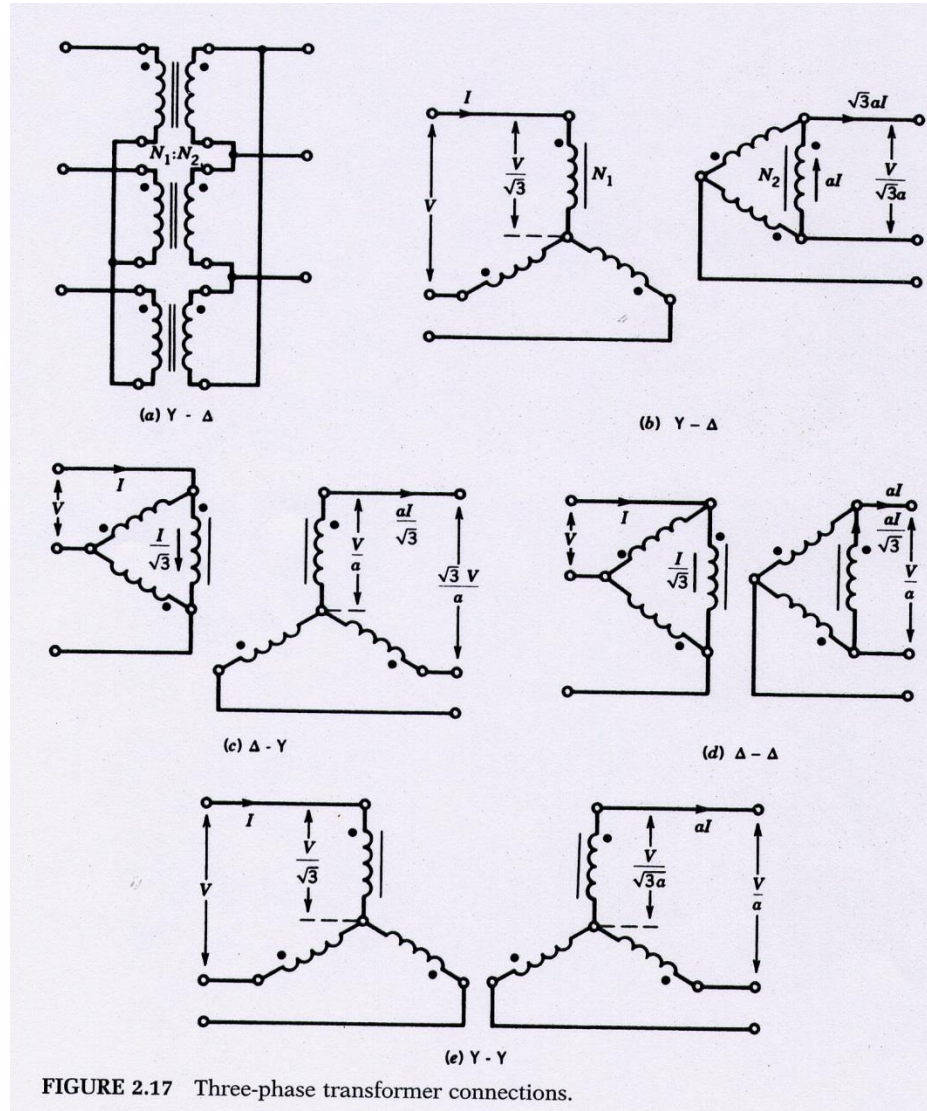
Star (Y) connection

- Line current is same as phase current
- Line-Line voltage is $\sqrt{3}$ phase-neutral voltage
- Power is given by $\sqrt{3} V_{L-L} I_L \cos\theta$ or $3V_{ph} I_{ph} \cos\theta$

Delta (Δ) connection

- Line-Line voltage is same as phase voltage
- Line current is $\sqrt{3}$ phase current
- Power is given by $\sqrt{3} V_{L-L} I_L \cos\theta$ or $3V_{ph} I_{ph} \cos\theta$

Typical three phase transformer connections



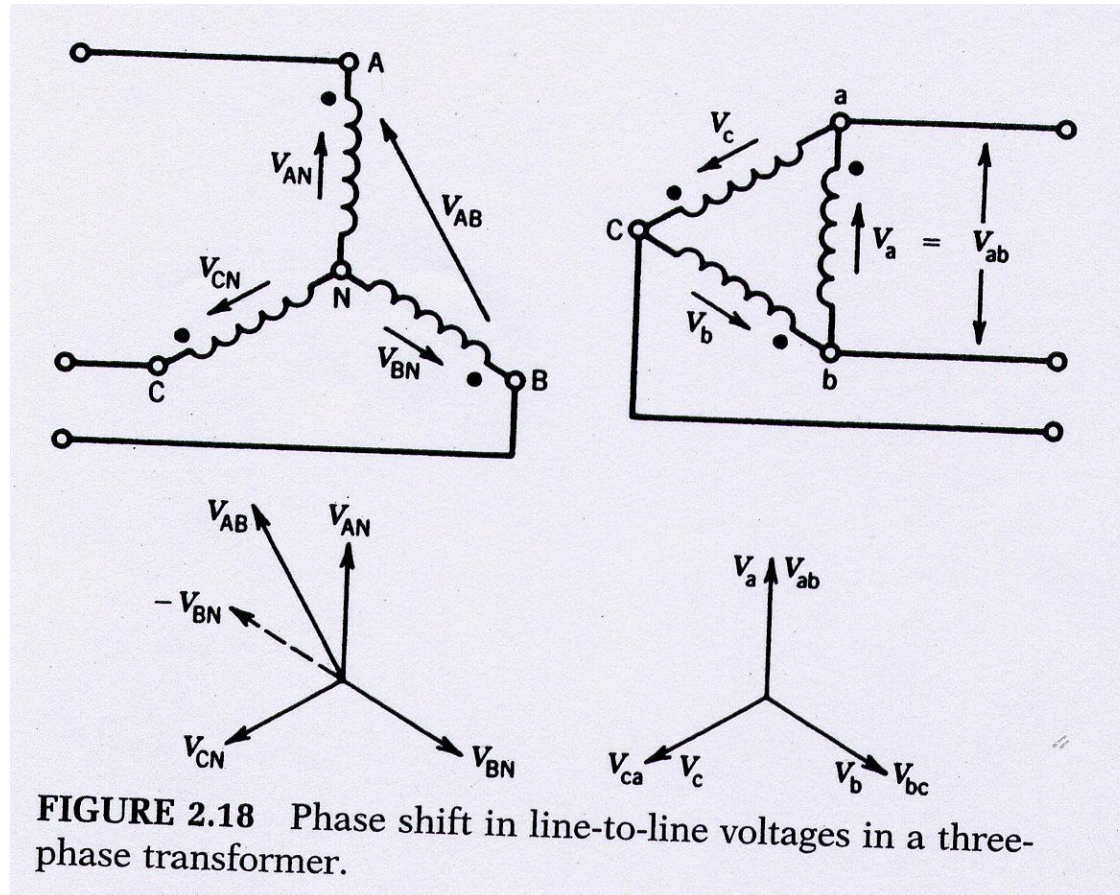
Other possible three phase transformer Connections

- Y- zigzag
- Δ - zigzag
- Open Delta or V
- Scott or T

How are three phase transformers made?

- Either by having three single phase transformers connected as three phase banks.
- Or by having coils mounted on a single core with multiple limbs
- The bank configuration is better from repair perspective, whereas the single three phase unit will cost less ,occupy less space, weighs less and is more efficient

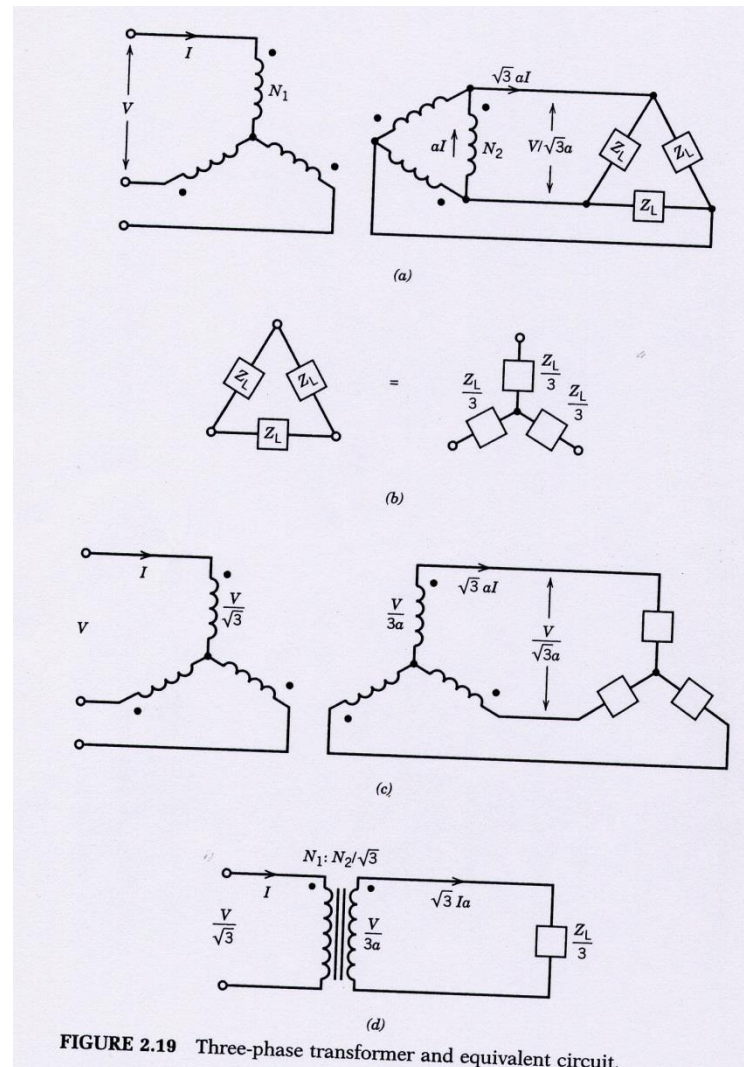
Phase-shift between line-line voltages in transformers



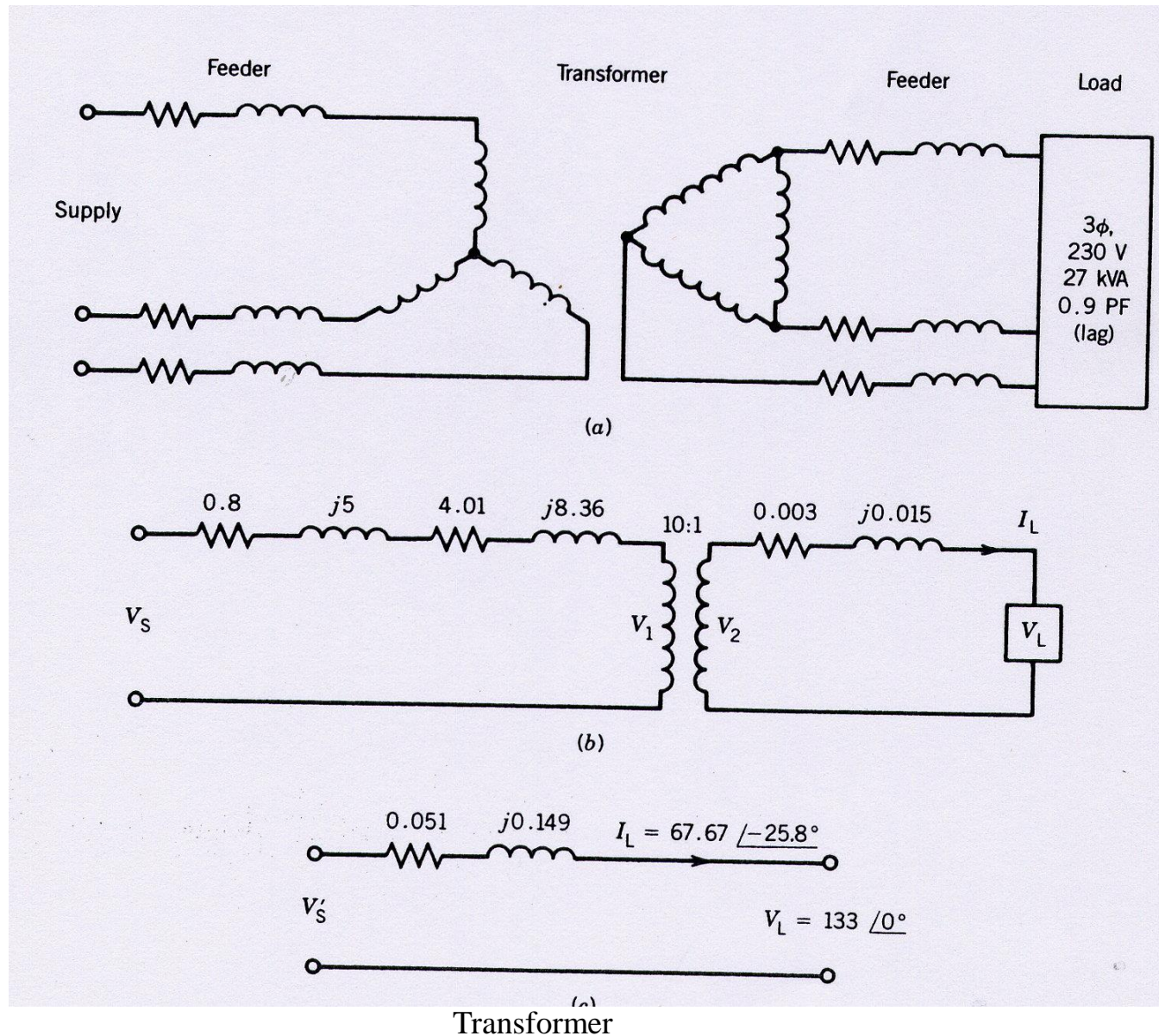
Vector grouping of transformers

- Depending upon the phase shift of line-neutral voltages between primary and secondary; transformers are grouped. This is done for ease of paralleling. Usually transformers between two different groups should not be paralleled.
- Group 1 : zero phase displacement (Yy0, Dd0, Dz0)
- Group 2 : 180° phase displacement (Yy6, Dd6, Dz6)
- Group 3 : 30° lag phase displacement (Dy1, Yd1, Yz1)
- Group 4 : 30° lead phase displacement (Dy11, Yd11, Yz11)
(Y=Y; D= Δ ; z=zigzag)

Calculation involving 3-ph transformers



An example involving 3-ph transformers



Open –delta or V connection

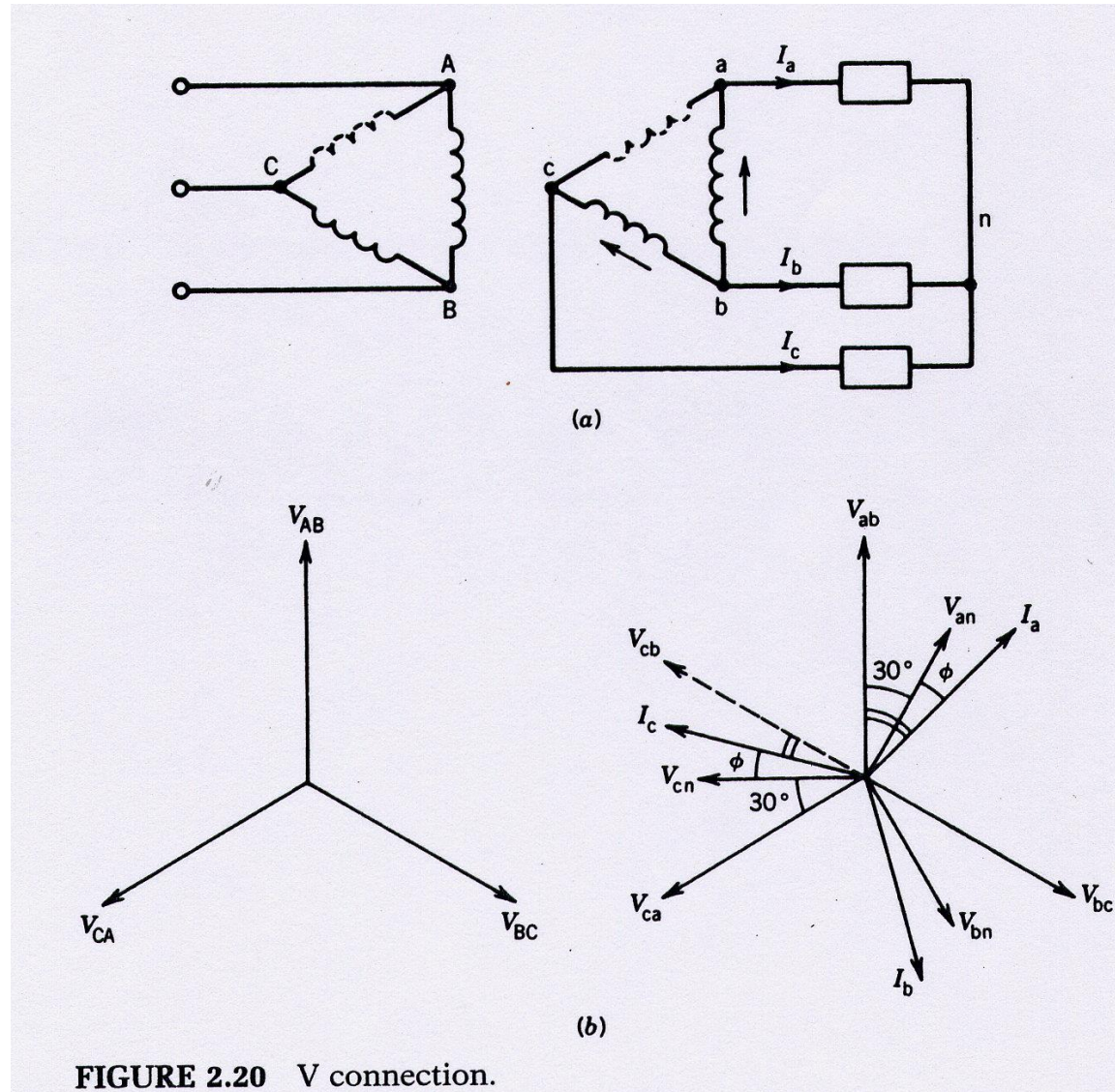


FIGURE 2.20 V connection.

Open –delta or V connection

Power from winding 'ab'

is $P_{ab} = V_{ab} I_a \cos(30^\circ + \phi)$

Power from winding 'bc'

is $P_{cb} = V_{cb} I_c \cos(30^\circ - \phi)$

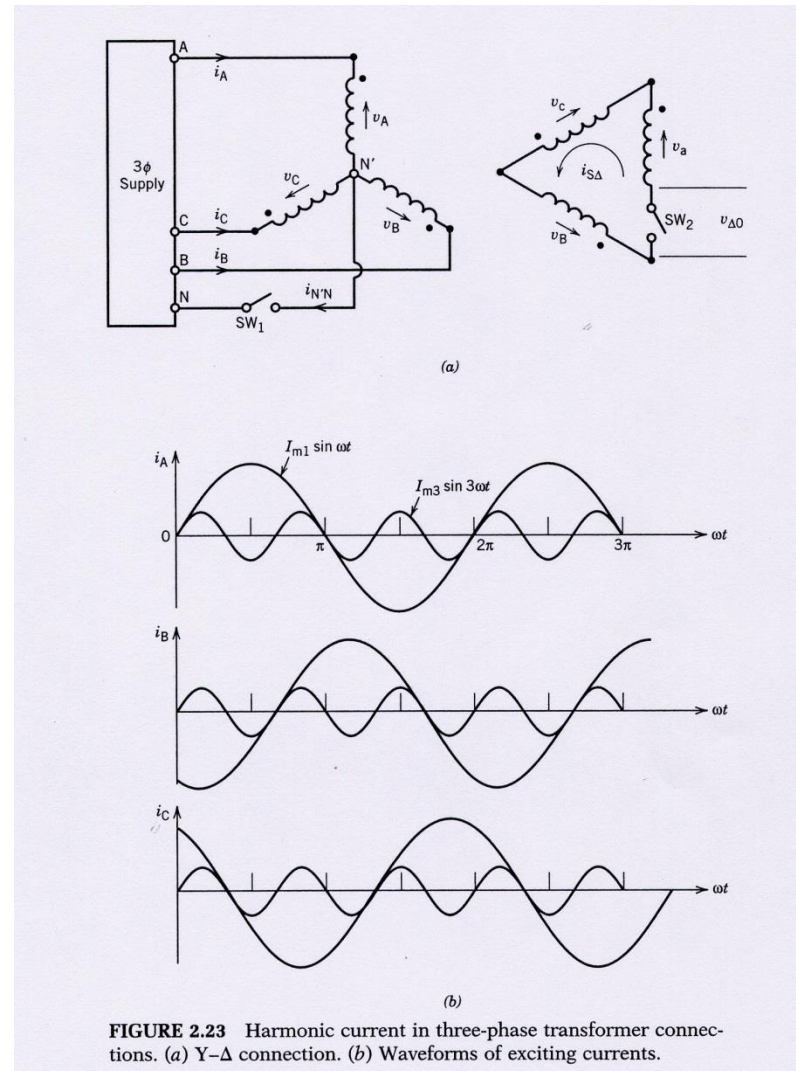
Therefore total power is

$= 2V_{L-L} I_L \cos 30^\circ \cos \phi$ or 57.7% of total
power from 3 phases

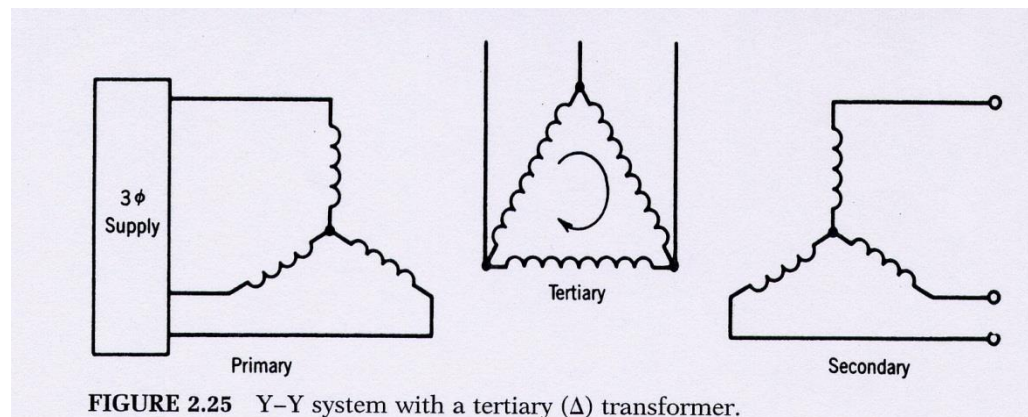
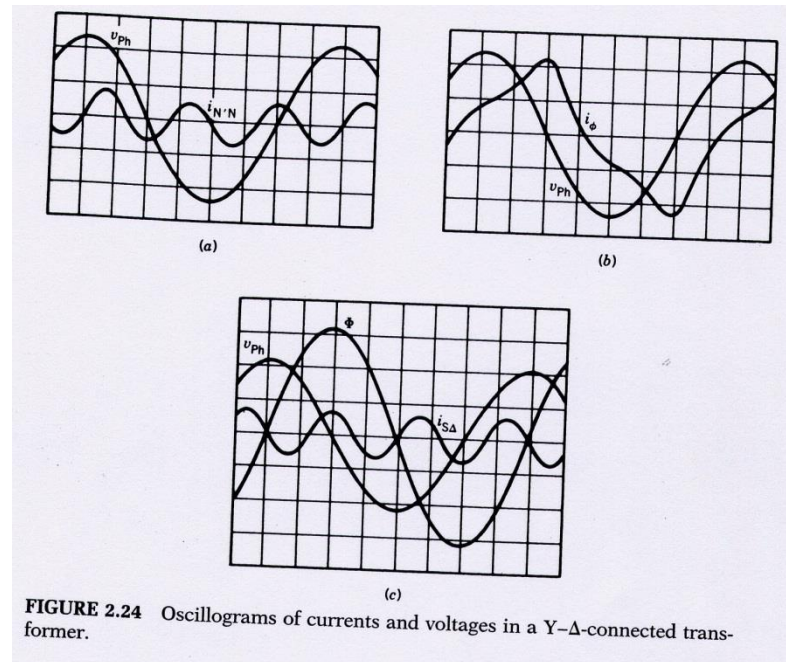
Harmonics in 3- ϕ Transformer Banks

- In absence of neutral connection in a Y-Y transformers 3rd harmonic current cannot flow
- This causes 3rd harmonic distortion in the phase voltages (both primary and secondary) but not line-line voltages, as 3rd harmonic voltages get cancelled out in line-line connections (see hw problem 2.22, where the voltage between the supply and primary neutrals is due to the third harmonic. This voltage can be modeled as a source in series with the fundamental voltage in the phase winding)
- Remedy is either of the following :
 - a) Neutral connections, b) Tertiary winding c) Use zigzag secondary d) Use star-delta or delta-delta type of transformers.
- a) The phenomenon is explained using a star-delta transformer.

Harmonics in 3- ϕ Transformer Banks(2)



Harmonics in 3- ϕ Transformer Banks(3)



Transformer

Per-Unit (pu) System

$$\bullet \text{Quantity in pu} = \frac{\text{actual value of quantity}}{\text{base value of quantity}}$$

- Values fall in a small zone and computational burden is less
- Easy to go from one side of a transformer to another without resorting to turns ratio multiplication and subsequent source of error
- Rated quantities (voltage,current,power) are selected as base quantities.
- Losses, regulation etc. can also be defined in pu.

Per-Unit (PU) System(2)

A single phase transformer is rated at 10kVA, 2200/220V, 60Hz. Equivalent impedance referred to high voltage side is $10.4 + j31.3 \Omega$. Find I_{base} , V_{base} , P_{base} , Z_{base} on both sides. What is the pu equivalent impedance on both sides? If magnetizing current I_m is 0.25 A on high voltage side what is its value in pu?

•HV side;

$$P_{\text{base}} = 10,000 \text{ VA} = 1 \text{ pu}, V_{\text{base}} = 2200 \text{ V} = 1 \text{ pu}$$

$$I_{\text{base}} = P_{\text{base}} / V_{\text{base}} = 4.55 \text{ A} = 1 \text{ pu}$$

$$Z_{\text{base}} = V_{\text{base}} / I_{\text{base}} = 2200 / 4.55 = 483.52 \Omega = 1 \text{ pu}$$

$$Z_{\text{eq(pu)}} = Z_{\text{eq}} / Z_{\text{base}} = 10.4 + j31.3 / 483.52 = 0.0215 + j0.0647 \text{ pu}$$

$$I_{m(\text{pu})} = I_m / I_{\text{base}} = 0.25 / 4.55 = 0.055 \text{ pu}$$

Per-Unit (PU) System(3)

•LV side;

$$P_{\text{base}}=10,000\text{VA}=1 \text{ pu}, V_{\text{base}}=220\text{V}=1 \text{ pu}$$

$$I_{\text{base}}=P_{\text{base}}/V_{\text{base}}=45.5\text{A}=1 \text{ pu}$$

$$Z_{\text{base}}=V_{\text{base}}/I_{\text{base}}=220/45.5=4.84 \Omega=1 \text{ pu}$$

$$Z_{\text{eq(pu)}}=Z_{\text{eq}}/Z_{\text{base}}=0.104+j0.313/4.84=0.0215+j0.0647 \text{ pu}$$

$$I_{\text{m(pu)}}=I_{\text{m}}/I_{\text{base}}=2.5/45.5=0.055 \text{ pu}$$

Transformer Construction

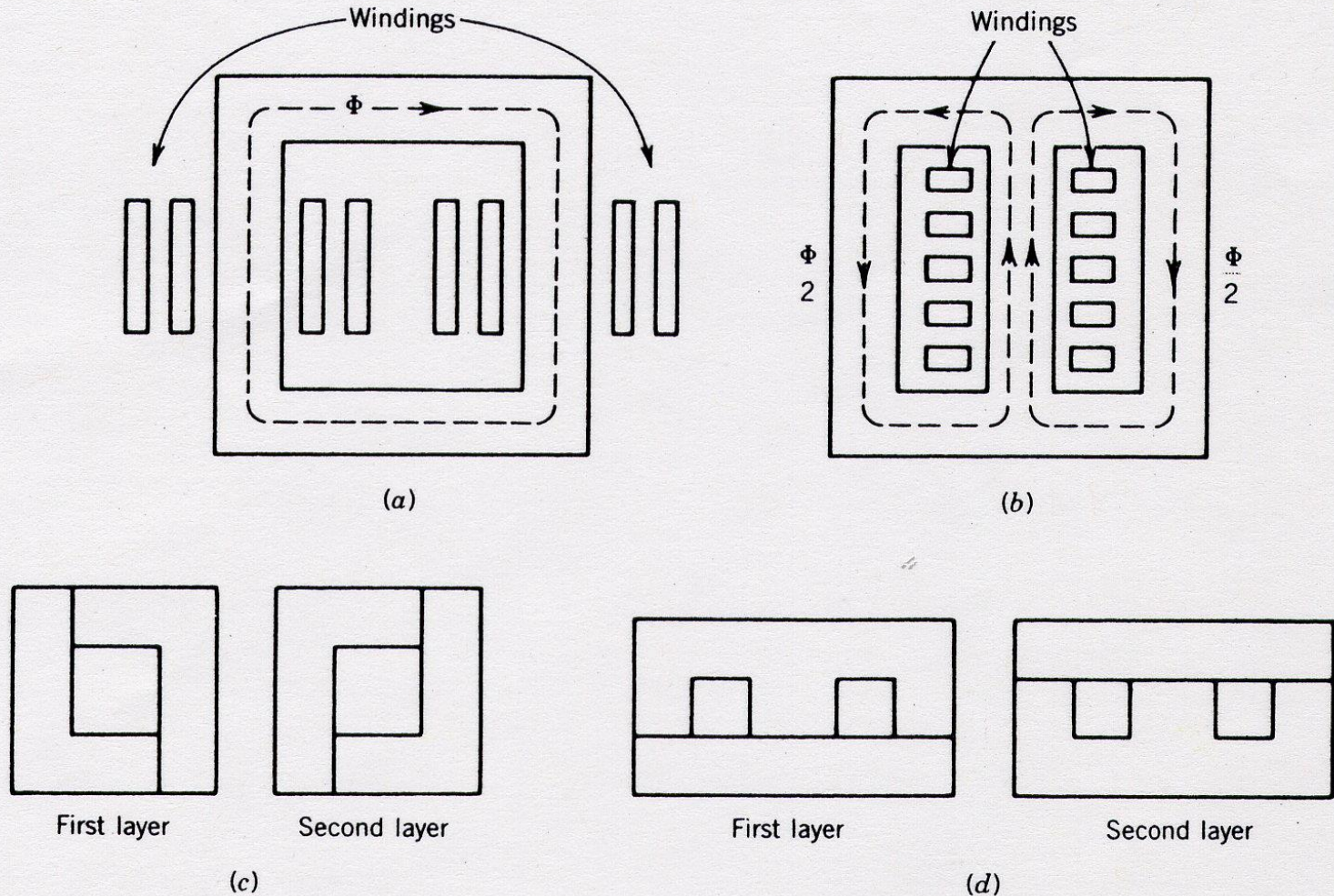


FIGURE 2.1 Transformer core construction. (a) Core-type, (b) Shell-type, (c) L-shaped lamination. (d) E-shaped lamination.

Transformer Construction(2)

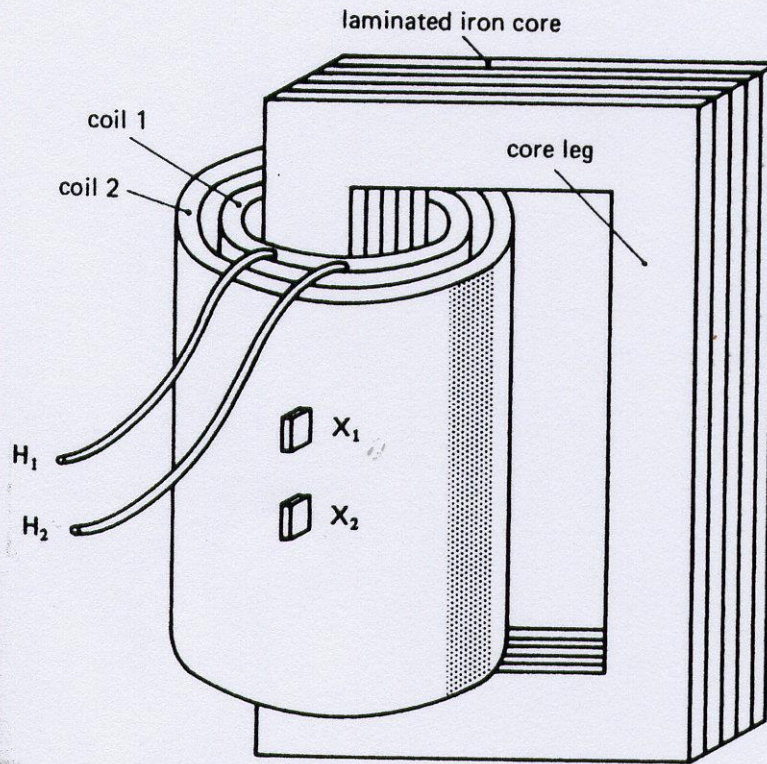


Figure 10.9a
Construction of a simple transformer.

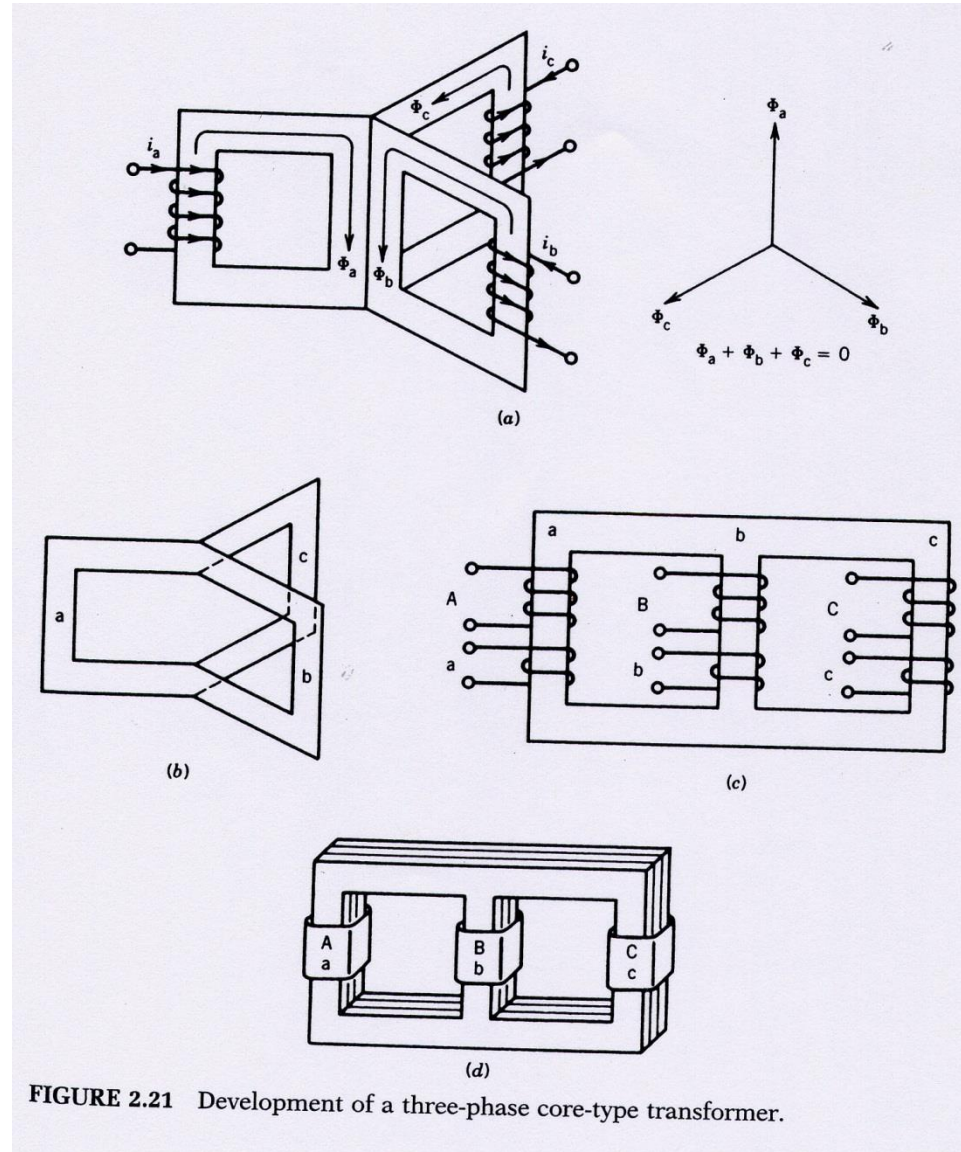


Figure 10.9b
Stacking laminations inside a coil.

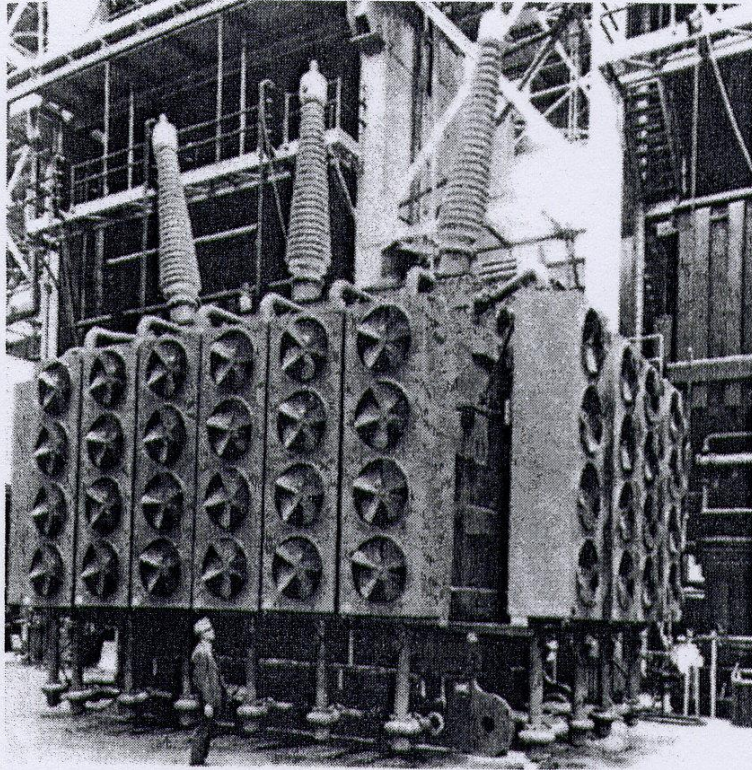
Left: Windings shown only on one leg

Right: Note the thin laminations

3- ϕ Transformer Construction (3)



3- ϕ Transformer Construction(4)



Left: A 1300 MVA, 24.5/345 kV, 60Hz transformer with forced oil and air (fan) cooling.

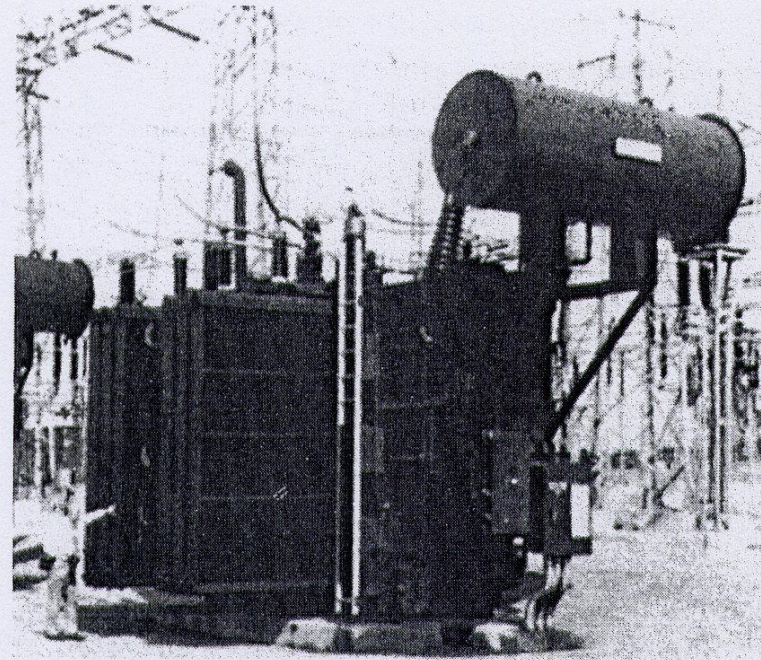


Figure 10.19
Three-phase, type OA/FA/FOA transformer rated

Right: A 60 MVA, 225/26.4 kV, 60 Hz showing the conservator.